LP38511-ADJ 800mA Fast-Transient Response Adjustable Low-Dropout Linear Voltage Regulator

Check for Samples: LP38511-ADJ

FEATURES

- **2.25V to 5.5V Input Voltage Range**
- **Adjustable Output Voltage Range of 0.5V to 3.3V**
- **800mA Output Load Current**
- **±2.0% Accuracy over Line, Load, and Full-Temperature Range from -40°C to +125°C**
- **Stable with Tiny 10 uF Ceramic Capacitors**
- **Enable Pin**
- **Typically Less than 1uA of Ground Pin Current with Enable Pin Low**
- **25dB of PSRR at 100 kHz**
- **Over-Temperature and Over-Current Protection**
- **8-Pin SO PowerPad and 5-Pin PFM Surface Mount Packages**

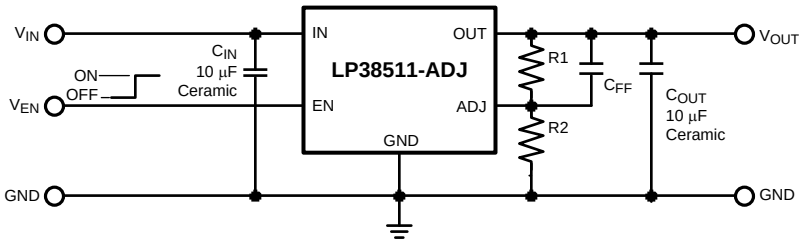
APPLICATIONS

- **Digital Core ASICs, FPGAs, and DSPs**
- **Servers**
- **Routers and Switches**
- **Base Stations**
- **Storage Area Networks**
- **DDR2 Memory**

DESCRIPTION

The LP38511-ADJ Fast-Transient Response Low-Dropout Voltage Regulator offers the highest-performance in meeting AC and DC accuracy requirements for powering Digital Cores. The LP38511-ADJ uses a proprietary control loop that enables extremely fast response to change in line conditions and load demands. Output Voltage DC accuracy is specified at 2.5% over line, load and full temperature range from -40°C to +125°C. The LP38511-ADJ is designed for inputs from the 2.5V, 3.3V, and 5.0V rail, is stable with 10 uF ceramic capacitors, and has an adjustable output voltage. The LP38511-ADJ provides excellent transient performance to meet the demand of high performance digital core ASICs, DSPs, and FPGAs found in highly-intensive applications such as servers, routers/switches, and base stations.

Typical Application Circuit



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Connection Diagram

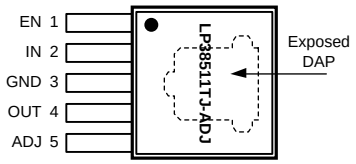

Figure 1. 5-Pin PFM, Top View
See NDQ0005A Package

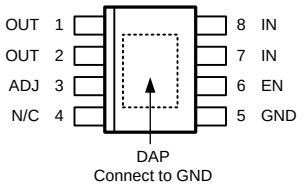

Figure 2. 8-Pin SO PowerPad, Top View
See DDA0008A Package

Pin Descriptions for PFM Package

Pin #	Pin Name	Function
1	EN	Enable. Pull high to enable the output, low to disable the output. This pin has no internal bias and must be tied to the input voltage, or actively driven.
2	IN	Input Supply Pin.
3	GND	Ground
4	OUT	Regulated Output Voltage Pin.
5	ADJ	The feedback to the internal Error Amplifier to set the output voltage.
DAP	DAP	The PFM DAP is used as a thermal connection to remove heat from the device to an external heat-sink in the form of the copper area on the printed circuit board. The DAP is physically connected to backside of the die. The DAP is internally connected to device ground. The DAP should be soldered to the Ground Plane copper.

Pin Descriptions for SO PowerPad Package

Pin #	Pin Name	Function
1, 2	OUT	Regulated Output Voltage Pins. Pins 1 and 2 share current and must be connected together.
3	ADJ	The feedback to the internal Error Amplifier to set the output voltage.
4	N/C	No internal connection
5	GND	Ground
6	EN	Enable. Pull high to enable the output, low to disable the output. This pin has no internal bias and must be tied to the input voltage, or actively driven.
7, 8	IN	Input Supply Pin. Pins 7 and 8 share current and must be connected together.
DAP	DAP	The SO PowerPad DAP is used as a thermal connection to remove heat from the device to an external heat-sink in the form of the copper area on the printed circuit board. The DAP is physically connected to backside of the die, but is not internally connected to device ground. The DAP should be soldered to the Ground Plane copper.

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

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Absolute Maximum Ratings[1][2]

Storage Temperature Range		−65°C to +150°C
Soldering Temperature[3]	PFM	260°C, 10s
	SO PowerPad	260°C, 10s
ESD Rating[4]		±2 kV
Power Dissipation[5]		Internally Limited
Input Pin Voltage (Survival)		-0.3V to +6.0V
Enable Pin Voltage (Survival)		-0.3V to +6.0V
Output Pin Voltage (Survival)		-0.3V to +6.0V
ADJ Pin Voltage (Survival)		-0.3V to +6.0V
I_{OUT} (Survival)		Internally Limited

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but does not ensure specific performance limits. For ensured specifications and conditions, see the Electrical Characteristics.
(2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
(3) Refer to JEDEC J-STD-020C for surface mount device (SMD) package reflow profiles and conditions. Unless otherwise stated, the temperatures and times are for Sn-Pb (STD) only.
(4) The human body model is a 100 pF capacitor discharged through a 1.5 kΩ resistor into each pin. Test method is per JESD22-A114.
(5) Device operation must be evaluated, and derated as needed, based on ambient temperature (T_A), power dissipation (P_D), maximum allowable operating junction temperature ($T_{J(MAX)}$), and package thermal resistance (θ_{JA}). The typical θ_{JA} ratings given are worst case based on minimum land area on two-layer PCB (EIA/JESD51-3). See POWER DISSIPATION/HEAT-SINKING for details.

Operating Ratings[1]

Input Supply Voltage, V_{IN}	2.25V to 5.5V
Output Voltage, V_{OUT}	V_{ADJ} to 5V
Enable Input Voltage, V_{EN}	0.0V to 5.5V
Output Current (DC)	1 mA to 800 mA
Junction Temperature[2]	−40°C to +125°C

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but does not ensure specific performance limits. For ensured specifications and conditions, see the Electrical Characteristics.
(2) Device operation must be evaluated, and derated as needed, based on ambient temperature (T_A), power dissipation (P_D), maximum allowable operating junction temperature ($T_{J(MAX)}$), and package thermal resistance (θ_{JA}). The typical θ_{JA} ratings given are worst case based on minimum land area on two-layer PCB (EIA/JESD51-3). See POWER DISSIPATION/HEAT-SINKING for details.

Electrical Characteristics

Unless otherwise specified: V_{IN}= 2.50V, V_{OUT}= V_{ADJ}, I_{OUT}= 10 mA, C_{IN}= 10 µF, C_{OUT}= 10 µF, V_{EN}= 2.0V. Limits in standard type are for T_J= 25°C only; limits in **boldface type** apply over the junction temperature (T_J) range of -40°C to +125°C. Minimum and Maximum limits are ensured through test, design, or statistical correlation. Typical values represent the most likely parametric norm at T_J= 25°C, and are provided for reference purposes only.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
V_{ADJ}	V_{ADJ} Accuracy[1]	2.25V ≤ V_{IN} ≤ 5.5V 10 mA ≤ I_{OUT} ≤ 800 mA	495.0 **490.0**	500.	505.0 **510.0**	mV
I_{ADJ}	ADJ Pin Bias Current	2.25V ≤ V_{IN} ≤ 5.5V	-	1	-	nA
$\Delta V_{ADJ}/\Delta V_{IN}$	V_{ADJ} Line Regulation[2][1]	2.25V ≤ V_{IN} ≤ 5.5V	-	0.03 **0.06**	-	%/V
$\Delta V_{ADJ}/\Delta I_{OUT}$	V_{ADJ} Load Regulation[3][1]	10 mA ≤ I_{OUT} ≤ 800 mA	-	0.10 **0.20**	-	%/A
V_{DO}	Dropout Voltage[4]	I_{OUT} = 800 mA	-	-	**260**	mV

(1) The line and load regulation specification contains only the typical number. However, the limits for line and load regulation are included in the output voltage tolerance specification.
(2) Line regulation is defined as the change in V_{ADJ} from the nominal value due to change in the voltage at the input.
(3) Load regulation is defined as the change in V_{ADJ} from the nominal value due to change in the load current at the output.
(4) Dropout voltage (V_{DO}) is typically defined as the input to output voltage differential (V_{IN} - V_{OUT}) where the input voltage is low enough to cause the output voltage to drop 2%. For the LP38511-ADJ, the minimum operating voltage of 2.25V is the limiting factor when the programed output voltage is less than typically 1.80V.

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Electrical Characteristics (continued)

Unless otherwise specified: V_{IN}= 2.50V, V_{OUT}= V_{ADJ}, I_{OUT}= 10 mA, C_{IN}= 10 µF, C_{OUT}= 10 µF, V_{EN}= 2.0V. Limits in standard type are for T_J= 25°C only; limits in **boldface type** apply over the junction temperature (T_J) range of -40°C to +125°C. Minimum and Maximum limits are ensured through test, design, or statistical correlation. Typical values represent the most likely parametric norm at T_J= 25°C, and are provided for reference purposes only.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
I_{GND}	Ground Pin Current, Output Enabled	I_{OUT} = 10 mA	-	7.5	11 **12**	mA
		I_{OUT} = 800 mA	-	9	11 **13**	
	Ground Pin Current, Output Disabled	V_{EN} = 0.50V	-	0.1	3.5 **12**	µA
I_{SC}	Short Circuit Current	V_{OUT} = 0V	-	1.5	-	A
Enable Input						
$V_{EN(ON)}$	Enable ON Voltage Threshold	V_{EN} rising from <0.5V until V_{OUT} = ON	0.90 **0.80**	1.20	1.50 **1.60**	V
$V_{EN(OFF)}$	Enable OFF Voltage Threshold	V_{EN} falling from 1.6V until V_{OUT} = OFF	0.60 **0.50**	1.00	1.40 **1.50**	V
$V_{EN(HYS)}$	Enable Voltage Hysteresis	$V_{EN(ON)}$ - $V_{EN(OFF)}$	-	200	-	mV
I_{EN}	Enable Pin Current	V_{EN} = V_{IN}	-	1	-	nA
		V_{EN} = 0V	-	-1	-	
$t_{d(OFF)}$	Turn-off delay	Time from V_{EN} < $V_{EN(OFF)}$ to V_{OUT} = OFF, I_{LOAD} = 800 mA	-	1	-	µs
$t_{d(ON)}$	Turn-on delay	Time from V_{EN} >$V_{EN(ON)}$ to V_{OUT} = ON, I_{LOAD} = 800 mA	-	25	-	
AC Parameters						
PSRR	Ripple Rejection	V_{IN} = 2.5V f = 120Hz	-	73	-	dB
		V_{IN} = 2.5V f = 1 kHz	-	70	-	
$\rho_{n(l/f)}$	Output Noise Density	f = 120Hz	-	0.4	-	µV/√Hz
e_n	Output Noise Voltage	BW = 10Hz - 100kHz	-	25	-	μV_{RMS}
Thermal Characteristics						
T_{SD}	Thermal Shutdown	T_J rising	-	165	-	°C
ΔT_{SD}	Thermal Shutdown Hysteresis	T_J falling from T_{SD}	-	10	-	
θ_{J-A}	Thermal Resistance Junction to Ambient(5)	SO PowerPad	-	168	-	°C/W
		PFM	-	67	-	
θ_{J-C}	Thermal Resistance Junction to Case	SO PowerPad	-	11	-	°C/W
		PFM	-	2	-	

(5) Device operation must be evaluated, and derated as needed, based on ambient temperature (T_A), power dissipation (P_D), maximum allowable operating junction temperature ($T_{J(MAX)}$), and package thermal resistance (θ_{JA}). The typical θ_{JA} ratings given are worst case based on minimum land area on two-layer PCB (EIA/JESD51-3). See POWER DISSIPATION/HEAT-SINKING for details.

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Typical Performance Characteristics

Unless otherwise specified: T_J = 25°C, V_{IN} = 2.50V, V_{OUT}= V_{ADJ}, V_{EN} = 2.0V, C_{IN} = 10 µF, C_{OUT} = 10 µF, I_{OUT} = 10 mA.

V_{ADJ} vs Temperature

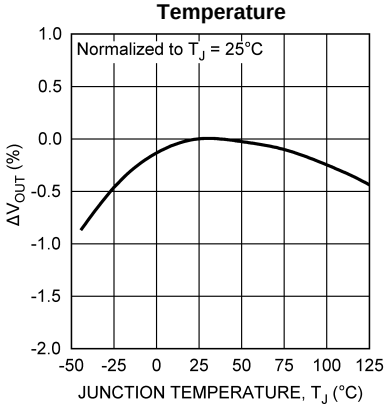

Figure 3.

V_{OUT} vs V_{IN}

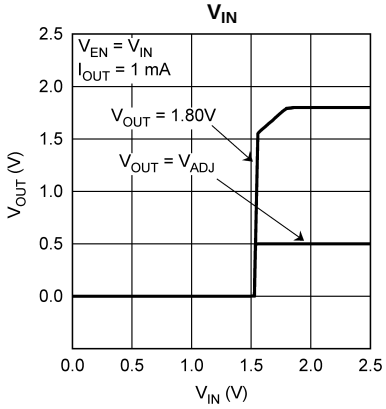

Figure 4.

Ground Pin Current (I_{GND}) vs V_{IN}

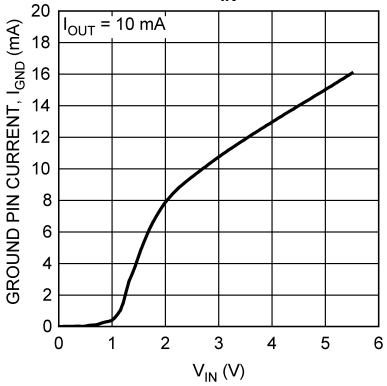

Figure 5.

Ground Pin Current (I_{GND}) vs Temperature

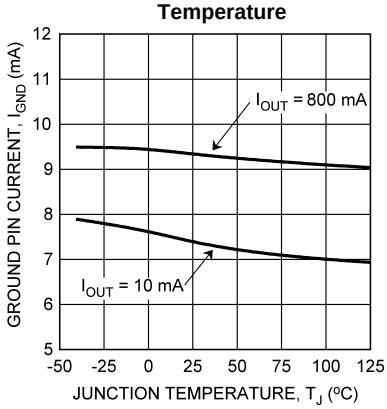

Figure 6.

Ground Pin Current (I_{GND}) vs Temperature

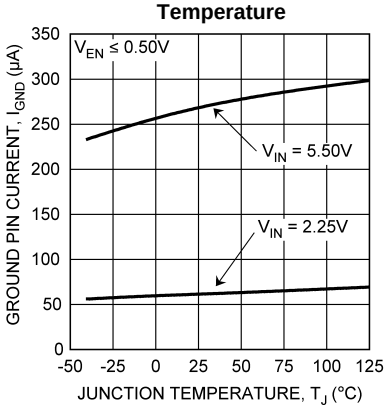

Figure 7.

Enable Threshold vs Temperature

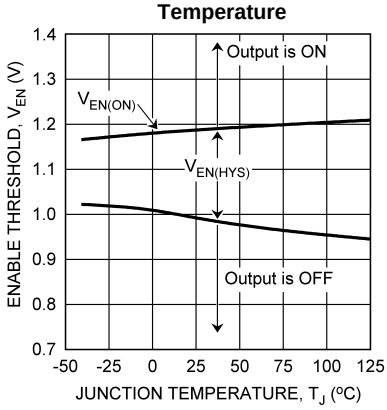

Figure 8.

Typical Performance Characteristics (continued)

Unless otherwise specified: T_J = 25°C, V_{IN} = 2.50V, V_{OUT}= V_{ADJ}, V_{EN} = 2.0V, C_{IN} = 10 µF, C_{OUT} = 10 µF, I_{OUT} = 10 mA.

V_{OUT} vs V_{EN}

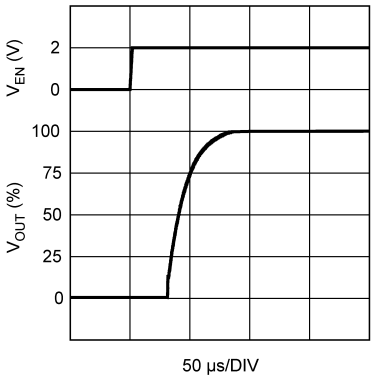

Figure 9.

Load regulation vs Temperature

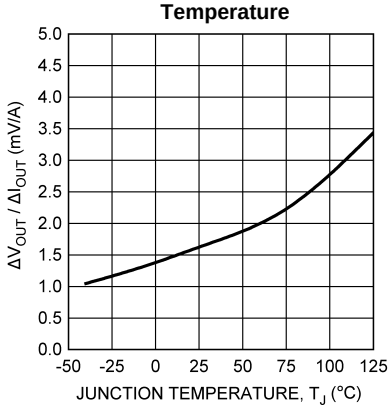

Figure 10.

Line Regulation vs Temperature

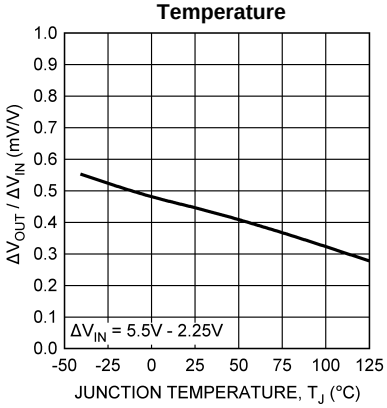

Figure 11.

Current Limit vs Temperature

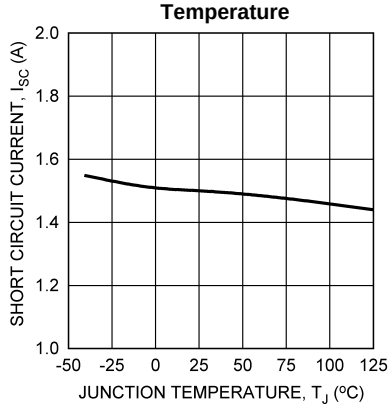

Figure 12.

Load Transient, 10 mA to 800 mA
V_{OUT} = V_{ADJ}, C_{OUT} = 10 µF Ceramic

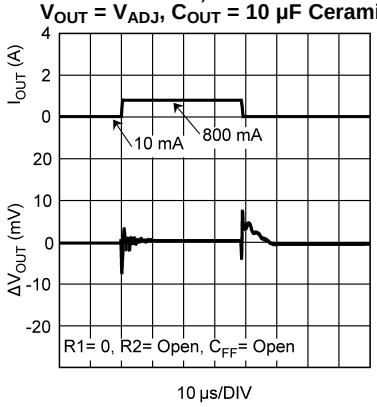

Figure 13.

Load Transient, 10 mA to 800 mA
V_{OUT} = 1.20V, C_{OUT} = 10 µF Ceramic

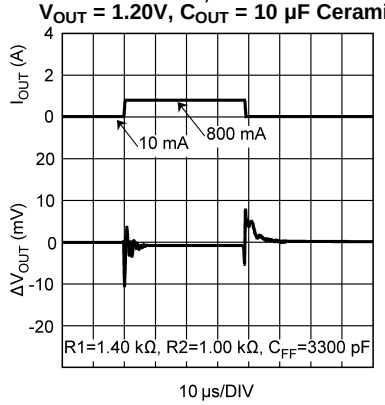

Figure 14.

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Typical Performance Characteristics (continued)

Unless otherwise specified: T_J = 25°C, V_{IN} = 2.50V, V_{OUT}= V_{ADJ}, V_{EN} = 2.0V, C_{IN} = 10 µF, C_{OUT} = 10 µF, I_{OUT} = 10 mA.

Load Transient, 250 mA to 800 mA
V_{OUT} = 1.20V, C_{OUT} = 10 µF Ceramic

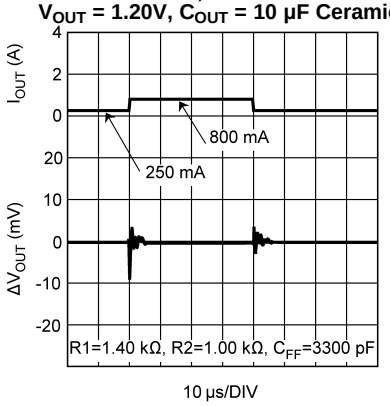


Figure 15.

Line Transient
V_{OUT} = V_{ADJ}, C_{OUT} = 10 µF Ceramic

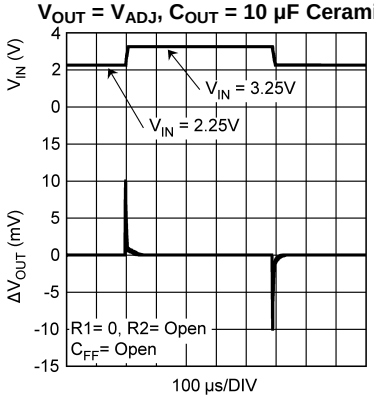


Figure 16.

Line Transient
V_{OUT} = 1.20V, C_{OUT} = 10 µF Ceramic

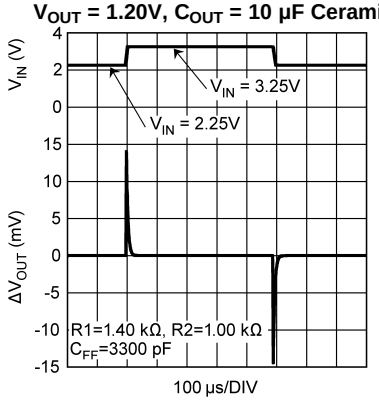


Figure 17.

PSRR, I_{OUT} = 100 mA
V_{OUT} = V_{ADJ}, C_{OUT} = 10 µF Ceramic

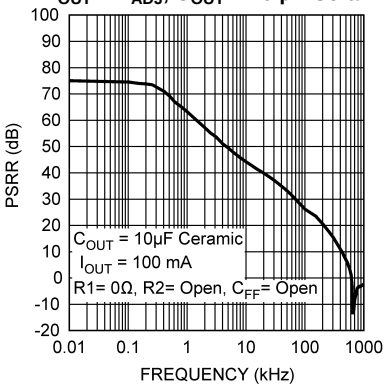


Figure 18.

PSRR, I_{OUT} = 800 mA
V_{OUT} = V_{ADJ}, C_{OUT} = 10µF Ceramic

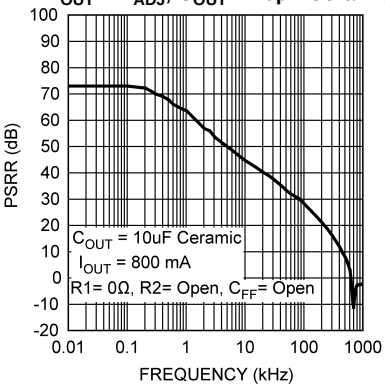


Figure 19.

Output Noise Density
V_{OUT} = V_{ADJ}, C_{OUT} = 10 µF Ceramic

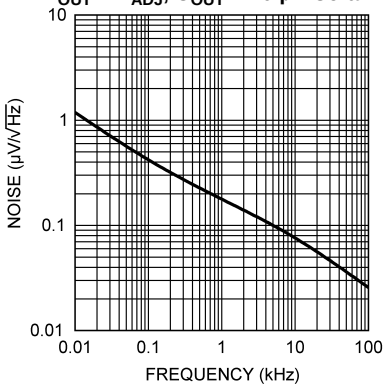
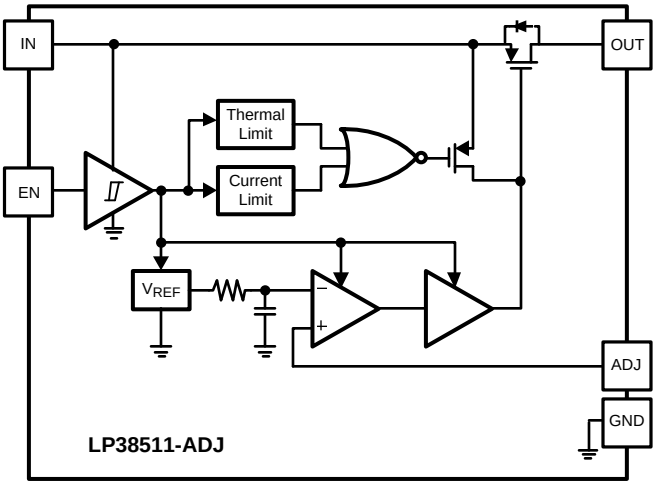


Figure 20.

Block Diagram



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APPLICATION INFORMATION

EXTERNAL CAPACITORS

Like any low-dropout regulator, external capacitors are required to assure stability. These capacitors must be correctly selected for proper performance.

Input Capacitor

A ceramic input capacitor of at least 10 µF is required. For general usage across all load currents and operating conditions, a 10 µF ceramic input capacitor will provide satisfactory performance.

Output Capacitor

A ceramic capacitor with a minimum value of 10 µF is required at the output pin for loop stability. It must be located less than 1 cm from the device and connected directly to the output and ground pin using traces which have no other currents flowing through them. As long as the minimum of 10 µF ceramic is met, there is no limitation on any additional capacitance.

X7R and X5R dielectric ceramic capacitors are strongly recommended, as they typically maintain a capacitance range within ±20% of nominal over full operating ratings of temperature and voltage. Of course, they are typically larger and more costly than Z5U/Y5U types for a given voltage and capacitance.

Z5U and Y5V dielectric ceramics are not recommended as the capacitance will drop severely with applied voltage. A typical Z5U or Y5V capacitor can lose 60% of its rated capacitance with half of the rated voltage applied to it. The Z5U and Y5V also exhibit a severe temperature effect, losing more than 50% of nominal capacitance at high and low limits of the temperature range.

REVERSE VOLTAGE

A reverse voltage condition will exist when the voltage at the output pin is higher than the voltage at the input pin. Typically this will happen when V_{IN} is abruptly taken low and C_{OUT} continues to hold a sufficient charge such that the input to output voltage becomes reversed. A less common condition is when an alternate voltage source is connected to the output.

There are two possible paths for current to flow from the output pin back to the input during a reverse voltage condition.

While V_{IN} is high enough to keep the control circuity alive, and the Enable pin is above the $V_{EN(ON)}$ threshold, the control circuitry will attempt to regulate the output voltage. Since the input voltage is less than the programmed output voltage, the control circuit will drive the gate of the pass element to the full on condition when the output voltage begins to fall. In this condition, reverse current will flow from the output pin to the input pin, limited only by the $R_{DS(ON)}$ of the pass element and the output to input voltage differential. Discharging an output capacitor up to 1000 µF in this manner will not damage the device as the current will rapidly decay. However, continuous reverse current should be avoided. When the Enable is low this condition will be prevented.

The internal PFET pass element in the LP38511-ADJ has an inherent parasitic diode. During normal operation, the input voltage is higher than the output voltage and the parasitic diode is reverse biased. However, if the output voltage to input voltage differential is more than 500 mV (typical) the parasitic diode becomes forward biased and current flows from the output pin to the input pin through the diode. The current in the parasitic diode should be limited to less than 1A continuous and 5A peak.

If used in a dual-supply system where the regulator output load is returned to a negative supply, the output pin must be diode clamped to ground. A Schottky diode is recommended for this protective clamp.

SHORT-CIRCUIT PROTECTION

The LP38511-ADJ is short circuit protected, and in the event of a peak over-current condition the short-circuit control loop will rapidly drive the output PMOS pass element off. Once the power pass element shuts down, the control loop will rapidly cycle the output on and off until the average power dissipation causes the thermal shutdown circuit to respond to servo the on/off cycling to a lower frequency. Please refer to the POWER DISSIPATION/HEAT-SINKING section for power dissipation calculations.

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SETTING THE OUTPUT VOLTAGE

The output voltage is set using the external resistive divider R1 and R2. The output voltage is given by the formula:

$$V_{OUT} = V_{ADJ} \times (1 + (R1/R2)) \quad (1)$$

The resistors used for R1 and R2 should be high quality, tight tolerance, and with matching temperature coefficients. It is important to remember that, although the value of V_{ADJ} is specified, the final value of V_{OUT} is not. The use of low quality resistors for R1 and R2 can easily produce a V_{OUT} value that is unacceptable.

It is recommended that the values selected for R1 and R2 are such that the parallel value is less than 1.00 kΩ. This is to reduce the possibility of any internal parasitic capacitances on the ADJ pin from creating an undesirable phase shift that may interfere with device stability.

$$((R1 \times R2) / (R1 + R2)) \leq 1.00\ k\Omega \quad (2)$$

FEED FORWARD CAPACITOR, C_{FF}

When using a ceramic capacitor for C_{OUT}, the typical ESR value will be too small to provide any meaningful positive phase compensation, F_Z, to offset the internal negative phase shifts in the gain loop.

$$F_Z = 1 / (2 \times \pi \times C_{OUT} \times ESR) \quad (3)$$

A capacitor placed across the gain resistor R1 will provide additional phase margin to improve load transient response of the device. This capacitor, C_{FF}, in parallel with R1, will form a zero in the loop response given by the formula:

$$F_Z = 1 / (2 \times \pi \times C_{FF} \times R1) \quad (4)$$

For optimum load transient response select C_{FF} so the zero frequency, F_Z, falls between 20 kHz and 40 kHz.

$$C_{FF} = 1 / (2 \times \pi \times R1 \times F_Z) \quad (5)$$

The phase lead provided by C_{FF} diminishes as the DC gain approaches unity, or V_{OUT} approaches V_{ADJ}. This is because C_{FF} also forms a pole with a frequency of:

$$F_P = 1 / (2 \times \pi \times C_{FF} \times (R1 \parallel R2)) \quad (6)$$

It's important to note that at higher output voltages, where R1 is much larger than R2, the pole and zero are far apart in frequency. At lower output voltages the frequency of the pole and the zero mover closer together. The phase lead provided from C_{FF} diminishes quickly as the output voltage is reduced, and has no effect when $V_{OUT} = V_{ADJ}$. For this reason, relying on this compensation technique alone is adequate only for higher output voltages.

Table 1 lists some suggested, best fit, standard ±1% resistor values for R1 and R2, and a standard ±10% capacitor values for C_{FF}, for a range of V_{OUT} values. Other values of R1, R2, and C_{FF} are available that will give similar results.

Table 1.

V_{OUT}	R1	R2	C_{FF}	F_Z
0.80V	1.07 kΩ	1.78 kΩ	4700 pF	31.6 kHz
1.00V	1.00 kΩ	1.00 kΩ	4700 pF	33.8 kHz
1.20V	1.40 kΩ	1.00 kΩ	3300 pF	34.4 kHz
1.50V	2.00 kΩ	1.00 kΩ	2700 pF	29.5 kHz
1.80V	2.94 kΩ	1.13 kΩ	1500 pF	36.1 kHz
2.00V	1.02 kΩ	340Ω	4700 pF	33.2 kHz
2.50V	1.02 kΩ	255Ω	4700 pF	33.2 kHz
3.00V	1.00 kΩ	200Ω	4700 pF	33.8 kHz
3.30V	2.00 kΩ	357Ω	2700 pF	29.5 kHz

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Please refer to Application Note *AN-1378 Method For Calculating Output Voltage Tolerances in Adjustable Regulators* SNVA112 for additional information on how resistor tolerances affect the calculated V_{OUT} value.

ENABLE OPERATION

The Enable ON threshold is typically 1.2V, and the OFF threshold is typically 1.0V. To ensure reliable operation the Enable pin voltage must rise above the maximum $V_{EN(ON)}$ threshold and must fall below the minimum $V_{EN(OFF)}$ threshold. The Enable threshold has typically 200 mV of hysteresis to improve noise immunity.

The Enable pin (EN) has no internal pull-up or pull-down to establish a default condition and, as a result, this pin must be terminated either actively or passively.

If the Enable pin is driven from a single ended device (such as the collector of a discrete transistor) a pull-up resistor to V_{IN}, or a pull-down resistor to ground, will be required for proper operation. A 1 kΩ to 100 kΩ resistor can be used as the pull-up or pull-down resistor to establish default condition for the EN pin. The resistor value selected should be appropriate to swamp out any leakage in the external single ended device, as well as any stray capacitance.

If the Enable pin is driven from a source that actively pulls high and low (such as a CMOS rail to rail comparator output), the pull-up, or pull-down, resistor is not required.

If the application does not require the Enable function, the pin should be connected directly to the adjacent V_{IN} pin.

POWER DISSIPATION/HEAT-SINKING

A heat-sink may be required depending on the maximum power dissipation ($P_{D(MAX)}$), maximum ambient temperature ($T_{A(MAX)}$)of the application, and the thermal resistance (θ_{JA}) of the package. Under all possible conditions, the junction temperature (T_J) must be within the range specified in the Operating Ratings. The total power dissipation of the device is given by:

$$P_D = ((V_{IN} - V_{OUT}) \times I_{OUT}) + ((V_{IN}) \times I_{GND}) \quad (7)$$

where I_{GND} is the operating ground current of the device (specified under Electrical Characteristics).

The maximum allowable junction temperature rise (ΔT_J) depends on the maximum expected ambient temperature ($T_{A(MAX)}$) of the application, and the maximum allowable junction temperature ($T_{J(MAX)}$):

$$\Delta T_J = T_{J(MAX)} - T_{A(MAX)} \quad (8)$$

The maximum allowable value for junction to ambient Thermal Resistance, θ_{JA}, can be calculated using the formula:

$$\theta_{JA} = \Delta T_J / P_{D(MAX)} \quad (9)$$

LP38511-ADJ is available in PFM and SO PowerPad surface mount packages. For a comparison of the PFM package to the standard TO-263 package see Application Note *AN-1797 PFM Package* (SNVA328). The thermal resistance depends on amount of copper area, or heat sink, and on air flow. See Application Note *AN-1520 A Guide to Board Layout for Best Thermal Resistance for Exposed Packages* (SNVA183) for guidelines.

Heat-Sinking the PFM Package

The DAP of the PFM package is soldered to the copper plane for heat sinking. The PFM package has a θ_{JA} rating of 67°C/W, and a θ_{JC} rating of 2°C/W. The θ_{JA} rating of 67°C/W includes the device DAP soldered to an area of 0.055 square inches (0.22 in x 0.25 in) of 1 ounce copper on a two sided PCB, with no airflow. See JEDEC standard EIA/JESD51-3 for more information.

Figure 21 shows a curve for the θ_{JA} of PFM package for different thermal via counts under the exposed DAP, using a four layer PCB for heat sinking. The thermal vias connect the copper area directly under the exposed DAP to the first internal copper plane only. See JEDEC standards EIA/JESD51-5 and EIA/JESD51-7 for more information.

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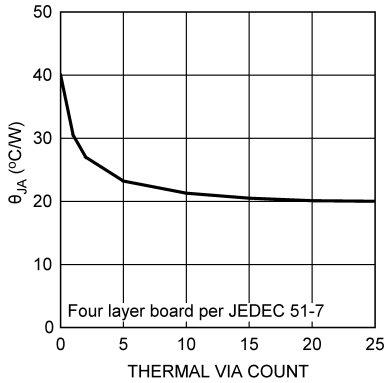

Figure 21. θ_{JA} vs Thermal Via Count for the PFM Package on 4–Layer PCB

Figure 22 shows the thermal performance when the PFM is mounted to a two layer PCB where the copper area is predominately directly under the exposed DAP. As shown in the figure, increasing the copper area beyond 1 square inch produces very little improvement.

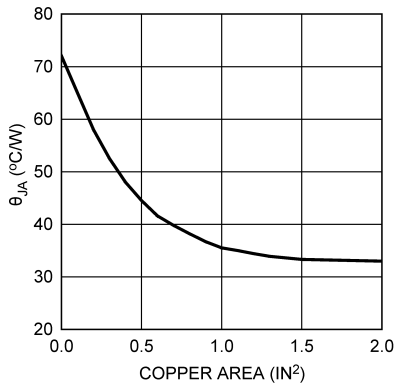

Figure 22. θ_{JA} vs Copper Area for the PFM Package

Heat-Sinking The SO PowerPad Package

The DAP of the SO PowerPad package is soldered to the copper plane for heat sinking. The LP38511MR package has a θ_{JA} rating of 168°C/W, and a θ_{JC} rating of 11°C/W. The θ_{JA} rating of 168°C/W includes the device DAP soldered to an area of 0.008 square inches (0.09 in x 0.09 in) of 1 ounce copper on a two sided PCB, with no airflow. See JEDEC standard EIA/JESD51-3 for more information.

Figure 23 shows a curve for different thermal via counts under the exposed DAP, using a four layer PCB for heat sinking. The thermal vias connect the copper area directly under the exposed DAP to the first internal copper plane only. See JEDEC standards EIA/JESD51-5 and EIA/JESD51-7 for more information.

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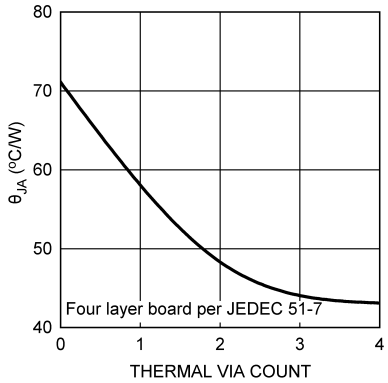

Figure 23. θ_{JA} vs Thermal Via Count for the SO PowerPad Package on 4–Layer PCB

Figure 24 shows thermal performance for a two layer board using thermal vias to a copper area on the bottom of the PCB. The copper area on the top of the PCB, which is soldered to the exposed DAP, is 0.10in x 0.20in, which is approximately the same dimensions as the body of the SO PowerPad package. The copper area on the bottom of the PCB is a square area and is centered directly under the SO PowerPad package.

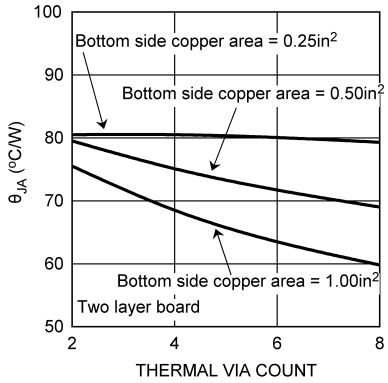

Figure 24. θ_{JA} vs Thermal Via Count for the SO PowerPad Package on 2–Layer PCB with Copper Area on Bottom-Side

Figure 25 shows thermal performance for a two layer board with the DAP soldered to copper area on the of the PCB only. Increasing the copper area soldered to the DAP to 1 square inch of 1 ounce copper, using a dog-bone type layout, will produce a typical θ_{JA} rating of 98°C/W.

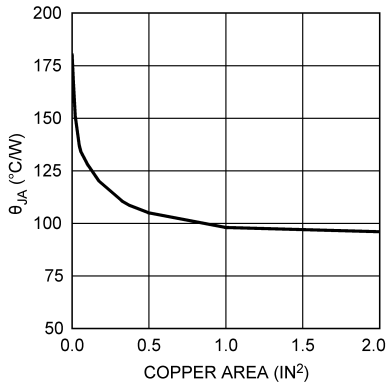

Figure 25. θ_{JA} vs Copper Area for the SO PowerPad Package on 2–Layer PCB with Copper Area on Top-Side

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REVISION HISTORY

Changes from Revision C (April 2013) to Revision D	Page
• Changed layout of National Data Sheet to TI format	13

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
LP38511MR-ADJ/NOPB	Active	Production	SO PowerPAD (DDA) | 8	95 | TUBE	Yes	SN	Level-3-260C-168 HR	-40 to 125	L38511 -ADJ
LP38511MR-ADJ/NOPB.A	Active	Production	SO PowerPAD (DDA) | 8	95 | TUBE	Yes	SN	Level-3-260C-168 HR	-40 to 125	L38511 -ADJ
LP38511MRX-ADJ/NO.A	Active	Production	SO PowerPAD (DDA) | 8	2500 | LARGE T&R	Yes	SN	Level-3-260C-168 HR	-40 to 125	L38511 -ADJ
LP38511MRX-ADJ/NOPB	Active	Production	SO PowerPAD (DDA) | 8	2500 | LARGE T&R	Yes	SN	Level-3-260C-168 HR	-40 to 125	L38511 -ADJ
LP38511TJ-ADJ/NOPB	Active	Production	TO-263 (NDQ) | 5	1000 | LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LP38511 TJ-ADJ
LP38511TJ-ADJ/NOPB.A	Active	Production	TO-263 (NDQ) | 5	1000 | LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LP38511 TJ-ADJ

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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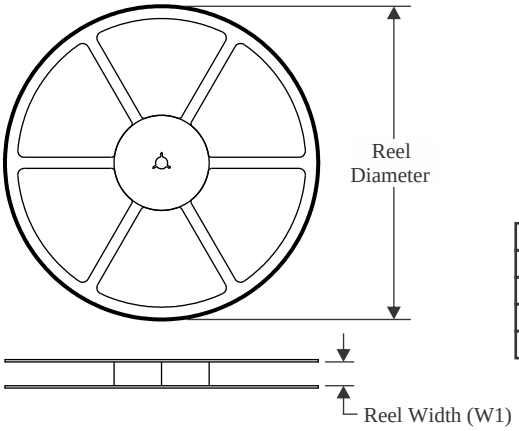
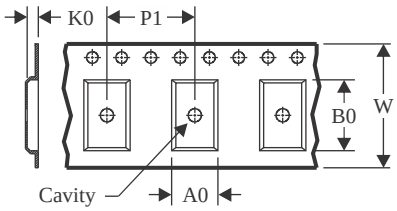
TAPE AND REEL INFORMATION

REEL DIMENSIONS

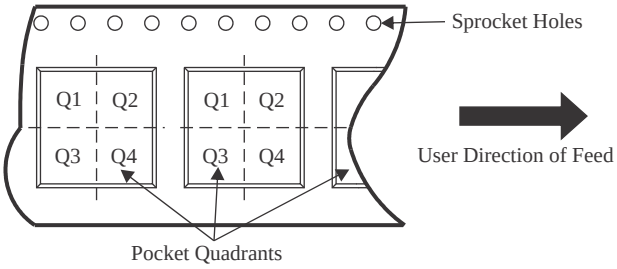


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

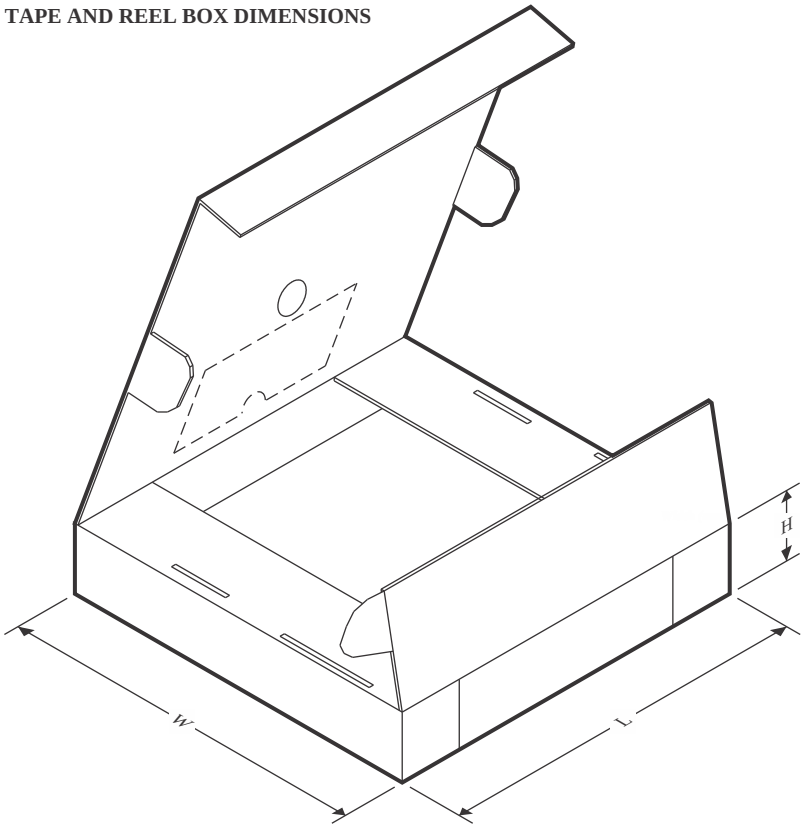
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LP38511MRX-ADJ/NOPB	SO PowerPAD	DDA	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LP38511TJ-ADJ/NOPB	TO-263	NDQ	5	1000	330.0	24.4	10.6	15.4	2.45	12.0	24.0	Q2

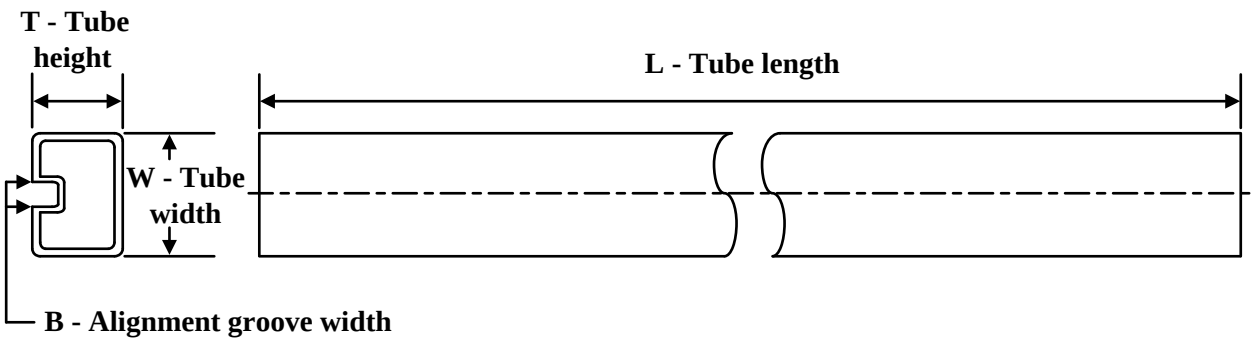
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LP38511MRX-ADJ/NOPB	SO PowerPAD	DDA	8	2500	356.0	356.0	36.0
LP38511TJ-ADJ/NOPB	TO-263	NDQ	5	1000	367.0	367.0	35.0

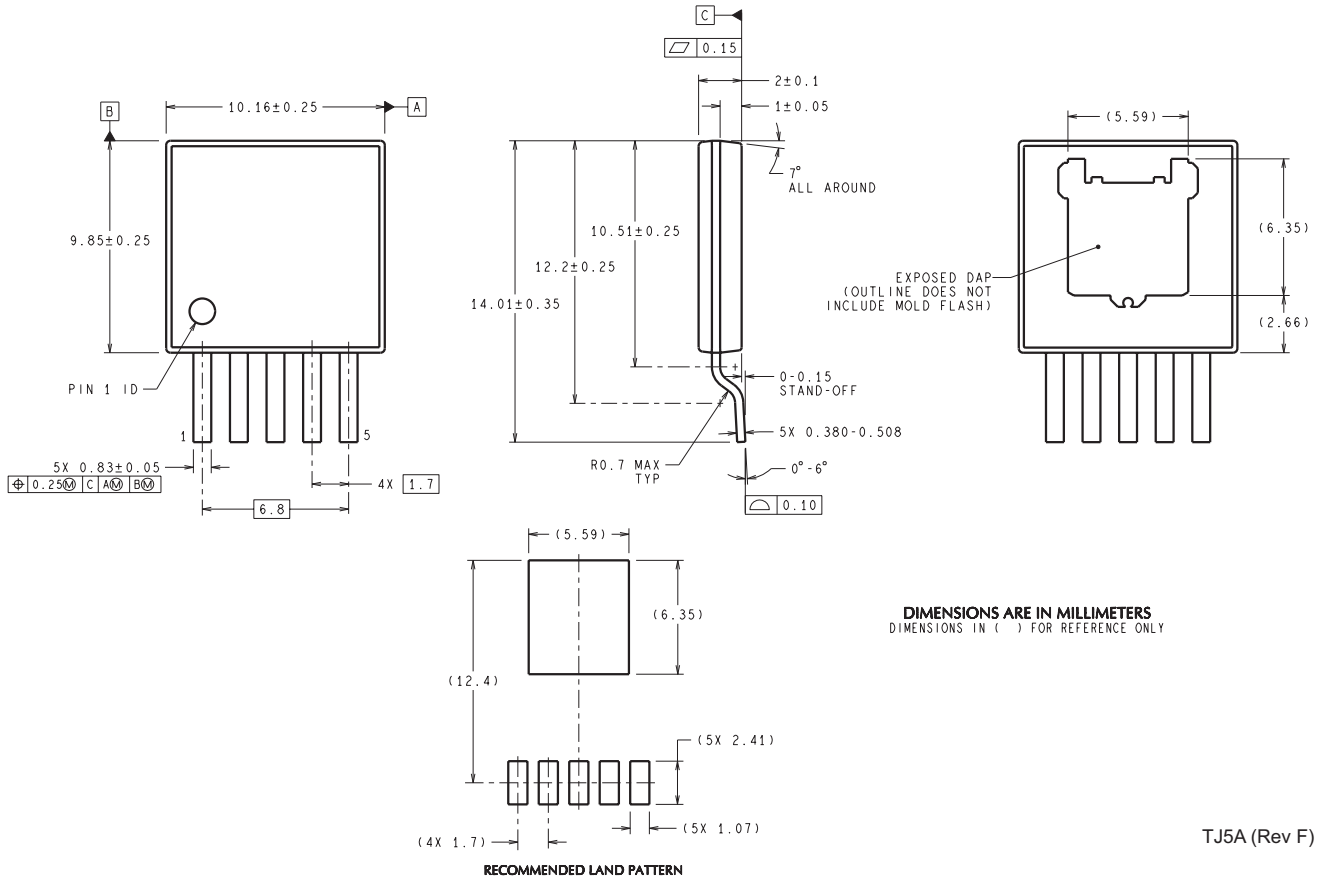
TUBE

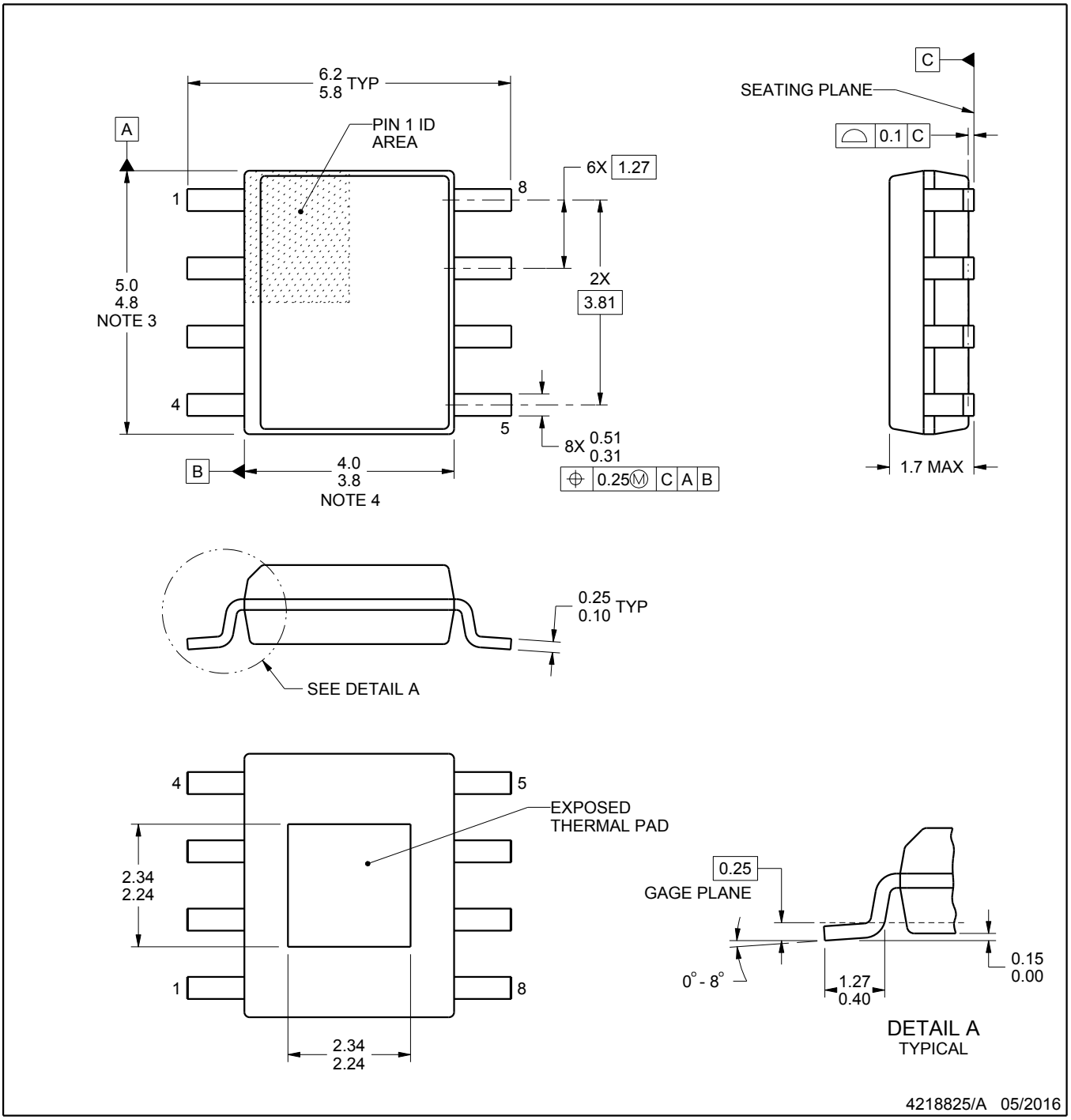


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
LP38511MR-ADJ/NOPB	DDA	HSOIC	8	95	495	8	4064	3.05
LP38511MR-ADJ/NOPB.A	DDA	HSOIC	8	95	495	8	4064	3.05

NDQ0005A


B
10.16±0.25
A
9.85±0.25
PIN 1 ID
1
5
5X 0.83±0.05
0.25 M C A M B M
4X 1.7
6.8
C
0.15
2±0.1
1±0.05
7°
ALL AROUND
10.51±0.25
12.2±0.25
14.01±0.35
0-0.15
STAND-OFF
5X 0.380-0.508
R0.7 MAX
TYP
0°-6°
0.10
(5.59)
(6.35)
EXPOSED DAP
(OUTLINE DOES NOT
INCLUDE MOLD FLASH)
(2.66)
(5.59)
(6.35)
(12.4)
(5X 2.41)
(4X 1.7)
(5X 1.07)
RECOMMENDED LAND PATTERN
DIMENSIONS ARE IN MILLIMETERS
DIMENSIONS IN ( ) FOR REFERENCE ONLY
TJ5A (Rev F)



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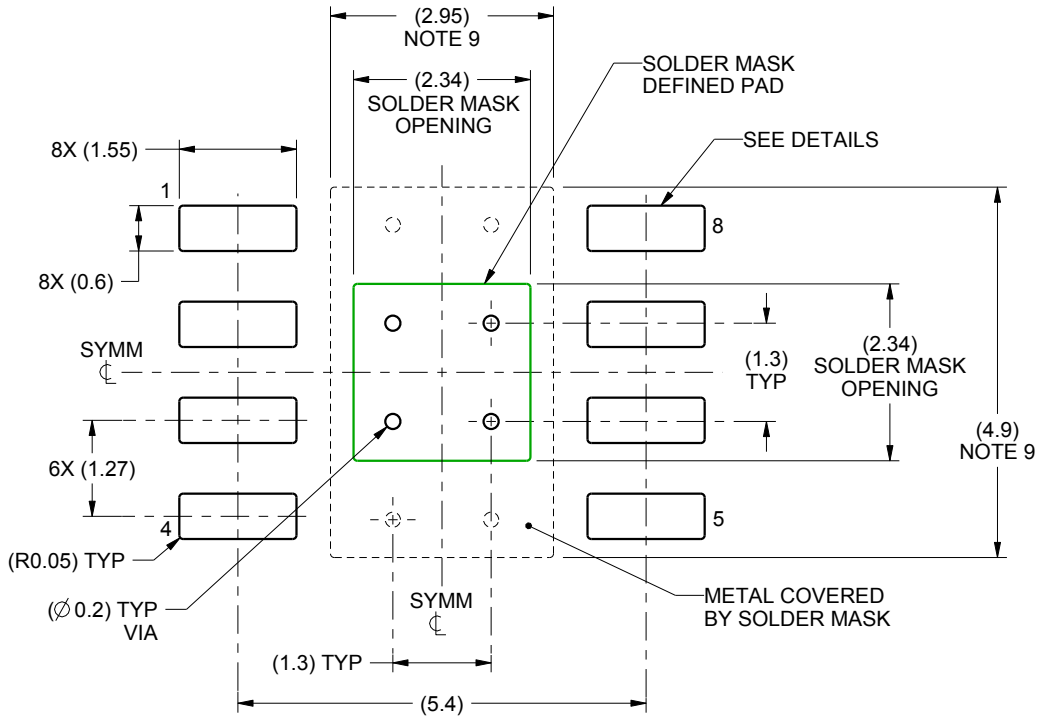
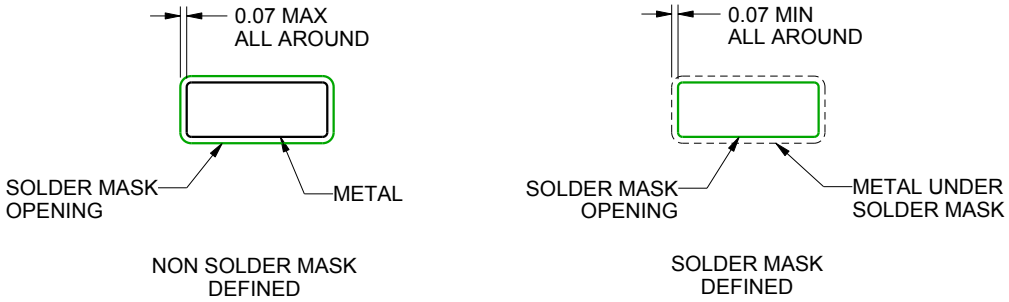
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MS-012.



LAND PATTERN EXAMPLE
SCALE:10X



SOLDER MASK DETAILS

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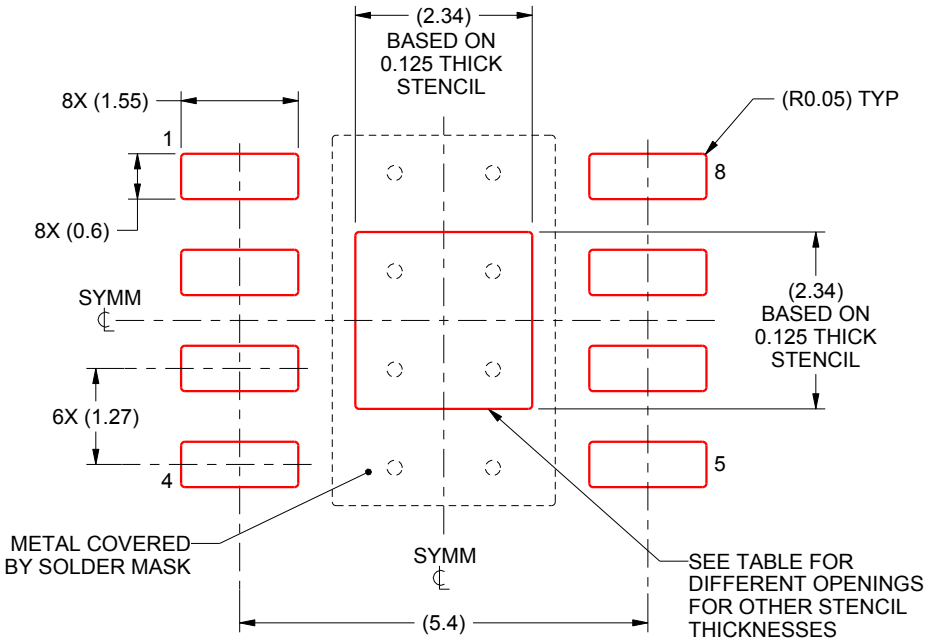
NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature numbers SLMA002 (www.ti.com/lit/slma002) and SLMA004 (www.ti.com/lit/slma004).
9. Size of metal pad may vary due to creepage requirement.
10. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



SOLDER PASTE EXAMPLE
EXPOSED PAD
100% PRINTED SOLDER COVERAGE BY AREA
SCALE:10X

STENCIL THICKNESS	SOLDER STENCIL OPENING
0.1	2.62 X 2.62
0.125	2.34 X 2.34 (SHOWN)
0.150	2.14 X 2.14
0.175	1.98 X 1.98

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NOTES: (continued)

11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
12. Board assembly site may have different recommendations for stencil design.

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